

# Toward true closed-loop neuromodulation: artifact-free recording during stimulation

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Closed-loop and responsive neuromodulation systems improve open-loop neurostimulation by responding directly to measured neural activity and providing adaptive, on-demand therapy. To be effective, these systems must be able to simultaneously record and stimulate neural activity, a task made difficult by persistent stimulation artifacts that distort and obscure underlying biomarkers. To enable simultaneous stimulation and recording, several techniques have been proposed. These techniques involve artifact-preventing system configurations, resilient recording front-ends, and back-end signal processing for removing recorded artifacts. Co-designing and integrating these artifact cancellation techniques will be key to enabling neuromodulation systems to stimulate and record at the same time. Here, we review the state-of-the-art for these techniques and their role in achieving artifact-free neuromodulation.

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**Current Opinion in Neurobiology** 2018, **50**:119–127

This review comes from a themed issue on **Neurotechnologies**

Edited by **Luo and Anikeeva**

<https://doi.org/10.1016/j.conb.2018.01.012>

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## Introduction

The past decade has seen a surge of clinical applications for neural recording and stimulation in the treatment of neurological disorders. For instance, responsive neurostimulation for treating epilepsy uses recorded neural biomarkers to trigger therapeutic stimulation [1–3]. Deep brain stimulation (DBS) is another therapy widely researched and clinically prescribed for the treatment of Parkinson's disease [4,5]. Closed-loop DBS has recently been proposed to only deliver stimulation when the patient is symptomatic, improving battery life and reducing side effects [6–8]. Looking forward, closed-loop therapies can be utilized to treat complex conditions

whose symptoms are not always present, necessitating simultaneous sensing, biomarker computation, and targeted therapeutic stimulation in the brain.

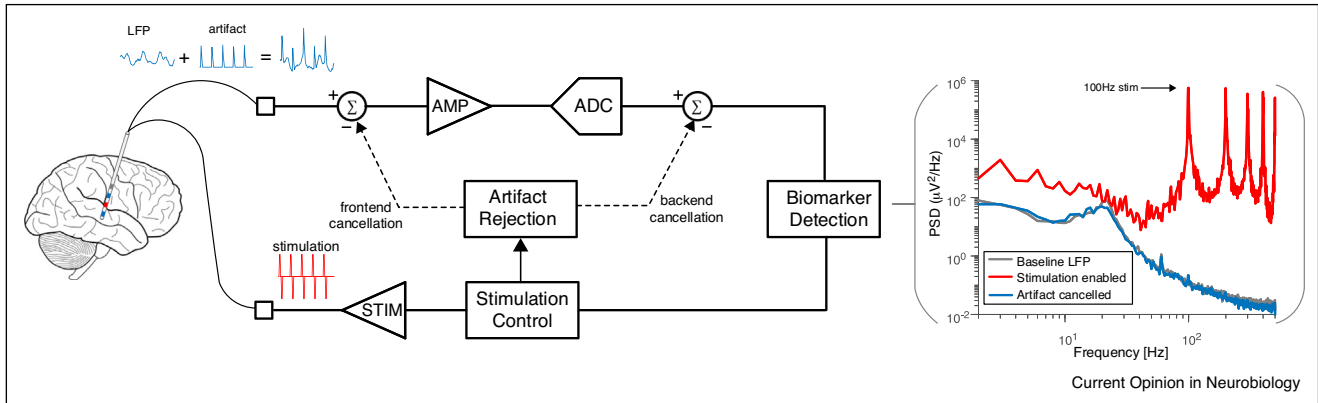
Today, there are major limitations in using therapeutic stimulation in a closed-loop, on-demand manner. Stimulation pulses create interference with recording electronics, which appear as artifacts masking the underlying neural signal and making simultaneous sensing and stimulating a challenge [9]. To record continuous neural activity while simultaneously delivering stimulation, neuromodulation systems must be resistant to large stimulation artifacts and be able to actively remove artifacts from the digitized signals for real-time biomarker computation (Figure 1). Here we review artifact mitigation techniques: artifact prevention, resilient recording front-ends for improved artifact immunity, and back-end digital signal processing for artifact removal. We conclude that system integration of multiple techniques produces the best results for enabling closed-loop artifact-free neuromodulation.

## Origin and prevention of stimulation artifacts

Stimulation artifacts typically consist of large voltage transients coinciding with the delivery of stimulation pulses. Their morphologies are dependent upon stimulator architecture and performance, stimulation waveform, and electrode configuration. Recorded artifacts consist of a short, high-amplitude peak (direct artifact) followed by a slow, exponential decay (residual artifact) superimposed on the local neural activity. Cancellation of these artifact components is crucial for analyzing spikes, as well as local field potentials (LFP) and electrocorticography (ECoG). Artifacts may be misclassified as spikes by some detection algorithms, and subsequent spikes cannot be recorded until the electrode voltage has returned to within the input range of the recording amplifier. The artifact peaks and ensuing decay together create strong distortions in the power spectrum at the stimulation frequency and also spreading into frequency bands of interest for LFP and ECoG.

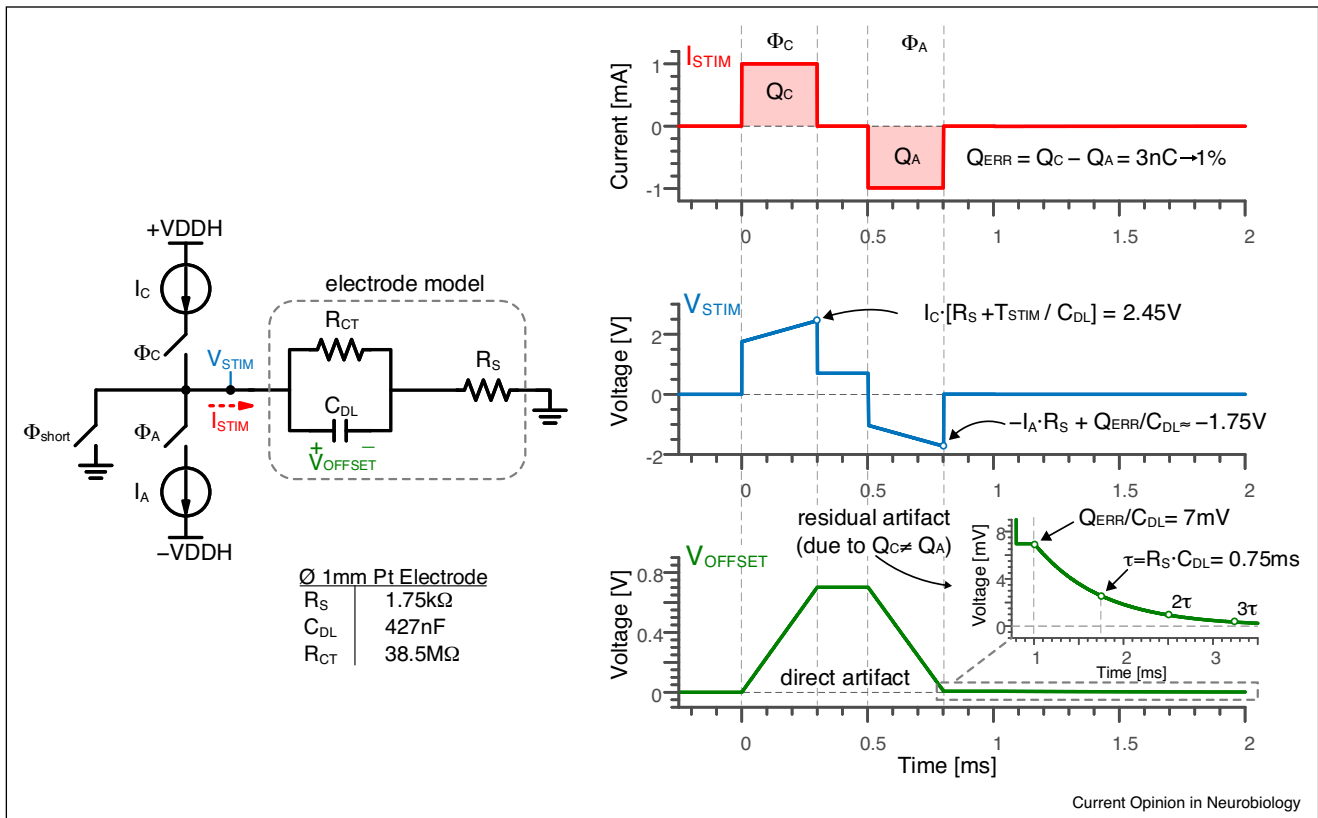
To demonstrate the origin of stimulation artifacts, a simplified electrical model of an electrode with biphasic current stimulation is shown in Figure 2. The electrode double-layer capacitance ( $C_{DL}$ ), charge-transfer resistance ( $R_{CT}$ ), and spread resistance ( $R_S$ ) are primarily functions of an electrode's area, geometry, surface roughness, and material [10]. Artifacts arise from voltage drops across stimulating electrodes and tissue as current passes

Figure 1



An example of adaptive closed-loop neuromodulation using local field potential feedback. Stimulation artifacts contaminate the recorded signal and prevent robust detection of biomarkers. Note that even an out-of-band stimulation signal can corrupt the entire sensing spectrum. To enable real-time feedback, neuromodulation systems attempt to mitigate artifact through the electrode configuration, resilient recording front-ends, and back-end cancellation methods.

Figure 2



Electrical model of the stimulation electrode and resulting current and voltage waveforms. Stimulation artifact,  $V_{OFFSET}$ , is a function of the stimulation current and electrode. Residual artifact duration, shown in the inset of  $V_{OFFSET}$ , depends on the accuracy of the charge balance and the time constant associated with dissipating residual charge stored on  $C_{DL}$ . Any mismatch in the stimulation phases (e.g. between  $I_C$  and  $I_A$ ) means charge will remain on the capacitors, resulting in a voltage of  $Q_{ERR}/C_{DL}$ . This residual charge will leak away slowly ( $\tau \sim R_S C_{DL}$ ). Front-ends must accommodate the large transient portion of the artifact and the slow discharge.

through them. This artifact will also propagate through the spread resistance to all of the other recording electrodes in the array. Precise modeling of the artifact is difficult since electrode impedance is nonlinear and varies with voltage. Furthermore, electrode impedance can change over time with chronic implantation [11]. The linear model, however, is effective for estimating the peak voltage due to stimulation and duration of an artifact [12\*,13,14].

Ideally, the cathodic and anodic phases of biphasic stimulation are perfectly matched with the same amount of charge ( $Q_C = Q_A$ ) and completely discharge the electrode capacitance,  $C_{DL}$ . In this case, the system would only experience direct artifact. However, any mismatch in current amplitude ( $I_C \neq I_A$ ) or pulse width ( $\Phi_C \neq \Phi_A$ ) leaves residual charge on the capacitors, resulting in DC voltage offset, a persisting artifact voltage that discharges slowly (Figure 2, residual artifact on  $V_{OFFSET}$ ). Residual artifact duration depends on the degree of mismatch and the time constant associated with dissipating residual charge. Aside from saturating recording front-ends, slow discharge can limit stimulation frequency since accumulated charge creates a DC current at the electrode, resulting in tissue damage and corrosion from electrolysis [10,15].

Several techniques have been utilized to improve stimulator phase matching. Pulse timing is typically well-controlled, whereas different current sources commonly have an amplitude mismatch of 1%, which is enough to produce significant artifact [12\*]. Some systems monitor  $V_{OFFSET}$  at the electrode from accumulated charge and calibrate the second phase current to minimize this offset [16]. Other systems calibrate prior to stimulation via current-copying using the cathodic current source as a reference for the anodic current [17,18]. Another alternative uses an H-bridge circuit, which is a stimulator topology that utilizes a single current source and a set of switches that allows the same current source to be used in both stimulation phases [19\*\*,20,21]. This method has been shown to achieve a mismatch between pulses of less than 0.02% [19\*\*].

Even stimulation pulses that are perfectly balanced with equal charge in each phase can generate large and long artifacts if recorded using inadequate circuits. Some work uses a tri-phasic stimulation waveform to minimize artifact duration [22]. Chu *et al.* [23] took an alternative approach by modeling the brain and interface as a communication channel and designing a waveform shape that inverts the transfer function, thereby reducing the artifact duration by 73%.

Deliberate placement of stimulation, recording, and reference electrodes has also been shown to reduce the stimulation artifact. For example, a symmetric

configuration between stimulation and recording electrodes can present the artifact as a common-mode signal, which can more easily be rejected by differential recording amplifiers [24\*\*]. Peterson *et al.* [25] eliminated a common ground between the recording and stimulation subsystems, allowing the recording reference to track common-mode artifacts. So far, the presented prevention techniques do not entirely eliminate artifacts, but do ease requirements on front-end acquisition.

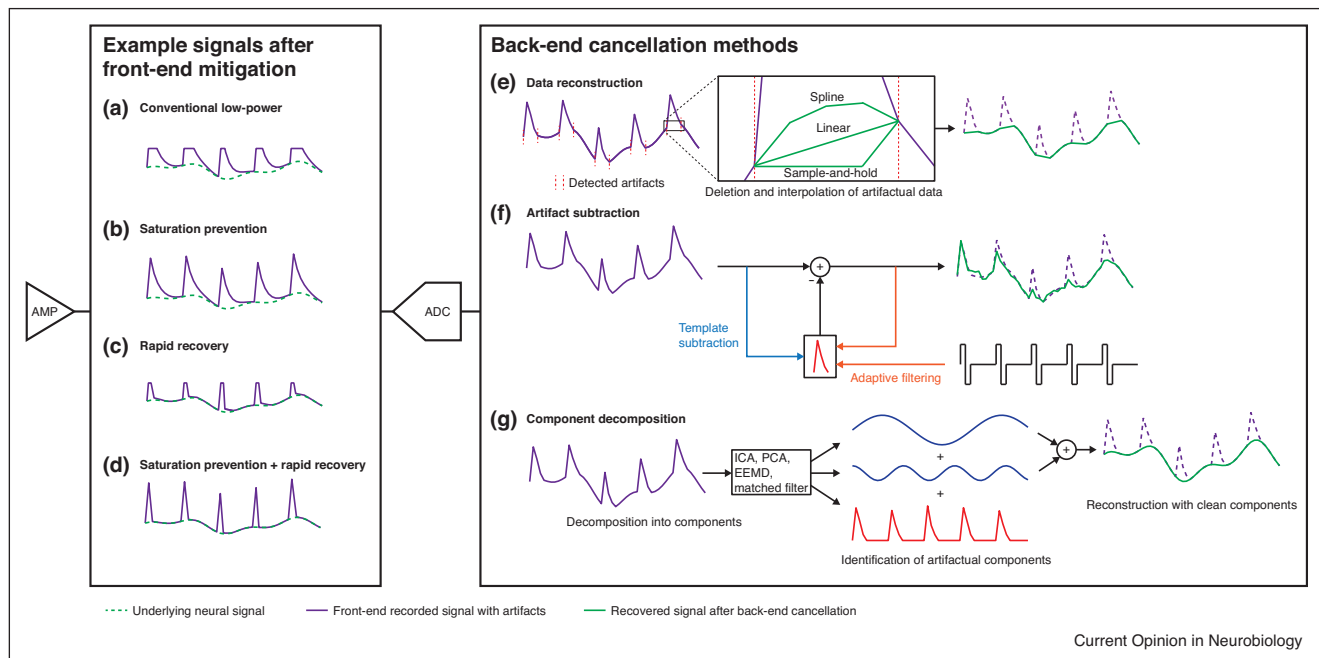
### Design for front-end artifact immunity

High fidelity recording of extracellular neural activity requires low-noise instrumentation to detect neural signals down to microvolt amplitudes. Noise performance trades off with power consumption, so recording front-ends, consisting of neural signal amplification and digitization circuitry, typically dominate the overall power of a neural recording system [26,27]. As a result, neural recording designs have primarily focused on optimizing power efficiency to extend battery life, reduce wireless power harvesting requirements, or minimize heat dissipation [28–31]. As recording systems scale to higher channel counts (>1000), power-efficient design is further exacerbated by tight area constraints [32–34] and high communication data rates [35].

Due to these constraints, most existing neural recording designs have not considered concurrent stimulation and are susceptible to artifacts. For example, power-efficient front-ends typically have a large gain to maximize their sensitivity. A high gain reduces the power requirements of subsequent processing stages, such as an analog-to-digital converter (ADC), but also causes the amplifier to saturate when presented with large stimulation artifacts. Furthermore, conventional front-ends utilize a low frequency high-pass corner to block DC offsets [29–31], but this results in a slow recovery from saturation due to the large, rapid transient voltage of a stimulation artifact (Figure 3a).

Newer front-end techniques have focused on mitigating the effects of stimulation artifacts by preventing saturation. One such technique is to increase the dynamic range of the recording circuits, enabling them to withstand and record larger voltages without significantly increasing the electronic noise (Figure 3b). A direct approach is to lower the amount of signal amplification [36]. A smaller artifact can remain in the linear range of the amplifier, so signal linearity is maintained and stimulation artifact can, in theory, be removed digitally in the back-end. However, this method sacrifices power efficiency by necessitating a higher supply voltage and a higher-resolution ADC for signal digitization. Another approach is to reduce the requirement for a large dynamic range by subtracting, using hardware methods, a model of the artifact at the input during stimulation to keep signals within a smaller range (Figure 3f). The model of the artifact is learned

Figure 3



Front-end artifact mitigation techniques improve linearity and duration of recorded artifacts, and back-end methods remove them. Examples of signals recorded by front-ends utilizing different mitigation methods are shown (left), followed by descriptions of back-end cancellation methods (right). Recorded analog signals (solid purple) contain artifacts that distort the underlying neural signal (dotted green) to various degrees. Back-end cancellation methods attempt to recover the neural signal (solid green). **(a)** With typical recording front-ends, artifacts saturate the amplifier and the amplifier recovers slowly. **(b)** Saturation can be prevented by increasing the dynamic range of the front-end, lowering signal distortion and enabling subtractive back-end methods. **(c)** By contrast, front-ends that quickly recover from a saturating signal can minimize the amount of distorted data. **(d)** Ideal systems implement both saturation prevention and rapid recovery, resulting in simple-to-cancel artifacts. **(e)** Simplest back-end techniques identify the artifact segments and reconstruct the underlying data using interpolation. **(f)** Subtractive methods remove artifact by subtracting estimated artifact waveforms. Estimation is done through template building or adaptive filtering. **(g)** Recorded waveforms can be separated into artifactual and neural components. Clean neural components are used to reconstruct the underlying signal.

through template building [32,37], filtering the stimulation pulse [38<sup>\*</sup>], or creating a replica artifact signal [39]. This technique holds promise, but can degrade the signal-to-noise ratio (SNR) of the underlying neural signal [38<sup>\*</sup>]. Alternatively, disconnecting the front-end via a series switch at the input prevents artifacts from reaching the recording circuitry [1,40–43]. However, this approach can suffer from slow transient settling once reconnected.

While spikes may be detected and analyzed even when superimposed on the long decays following amplifier saturation, these settling responses severely degrade LFP and ECoG signals. Thus, a resilient recording front-end must also rapidly recover from a saturating signal (Figure 3c). Resetting the recording circuits at every sample can clear the saturating charge and eliminate the long transient responses that result. Several designs add hard reset switches [19<sup>\*\*</sup>,44] or a controllable weak resistance [12<sup>\*</sup>,45,46] to reset the amplifier and rapidly return to the baseline voltage. In some cases, stimulation causes charge to accumulate on the recording electrode, so some work actively discharges the electrode itself to an averaged pre-stimulus voltage [12<sup>\*</sup>,13,14].

Johnson *et al.* [19<sup>\*\*</sup>] combines techniques by resetting at every sample while also increasing front-end dynamic range (Figure 3d). The design uses a mixed-signal approach that combines analog and digital domains, and places the ADC inside the amplifier feedback loop. This has several advantages, as it reduces normally saturating signal swings while providing the same overall gain, allowing for an increased input range of 100 s of mV while maintaining power and noise performance.

### Back-end neural signal recovery

Back-end signal processing techniques can be applied to the digitized signals at the ADC output to remove remaining stimulation artifacts. We divide these techniques into three categories: firstly, data reconstruction (Figure 3e), secondly, artifact subtraction (Figure 3f), and thirdly, component decomposition (Figure 3g). Although many of these techniques originated as offline methods, some have been implemented online for real-time artifact cancellation. Closed-loop and responsive applications require fast, low-complexity, and low-power online implementations.

Reconstruction methods remove samples contaminated with artifacts and replace them with interpolated values (Figure 3e). Sample-and-hold methods hold over the last known good sample for the duration of each artifact [47,48]. This requires only a single sample of memory, but may cause significant distortion. To reduce distortion, samples may be replaced by linear interpolation between the nearest clean samples [49\*,50\*], an estimation from a learned Gaussian probability density for data segments [51], or a reconstruction using cubic spline interpolation [52]. Hoffman *et al.* [51] report an SNR of between 30 and 40 dB for linear interpolation and Gaussian estimation, and Zhou *et al.* [49\*] show that baseline power spectral density for LFP could be accurately recovered after linear interpolation over artifacts.

Although simple to implement, data reconstruction requires artifact detection. This can be done using blind detection algorithms [48,50\*,51], or using timing indicators from the stimulator [19\*\*,37,49\*]. These methods lose information during the artifact, degrading SNR and potentially also removing action potentials which are typically shorter in duration than the artifact. Thus, interpolation is better suited to lower frequency recordings like LFP and ECoG than spike recordings. Ideally, this technique is paired with rapid recovery front-ends to keep the artifact short. High dynamic range is less critical as saturated data is discarded anyway.

High dynamic range front-ends, however, are essential for subtraction and component decomposition techniques, which require the undistorted artifact waveform. These algorithms assume that the artifact is linearly superimposed onto the neural signal and can be subtracted either sample-by-sample or as an interference source. Theoretically, decomposition and subtraction allow for the most accurate recovery of the underlying signal, crucial for spike detection and sorting.

First introduced as an offline method [48,53,54], template subtraction has been implemented in both online hardware [55\*] and software [37,56] (Figure 3f). Templates may be formed from averaging artifacts [37,48,53,54,55\*,56–58] or fitting artifacts to a predefined function type [9,53,59]. These subtraction techniques suffer from varying artifact morphology resulting from undersampling the artifact shape and misalignment of stimulation and sample timing [49\*,57]. To improve template accuracy, more complex methods resample and shift the artifact waveforms and templates [57,58]. Similarly, adaptive filtering estimates the artifact by filtering the stimulation pulse [38\*] or the artifact recorded on a neighboring channel [60], then subtracts it while filter coefficients are adapted. These subtraction methods can be implemented with low latency, but require artifact detection, template building and on-board memory for template storage. Templates must be updated often to

track any changes to artifact shape or stimulus waveform, otherwise signal distortion may occur. Often estimated templates take time to converge, resulting in varying levels of cancellation over time.

Component decomposition methods separate ensembles of recorded channels into artifact and non-artifact components and reconstruct a clean neural signal with only the non-artifact components (Figure 3g). Ensemble empirical mode decomposition [61,62] and independent component analysis [61,63] are common approaches to blindly separating artifacts from neural source. While these methods offer great accuracy in reconstruction, they also involve the most intensive computation, requiring iterative processing steps. Therefore, to our knowledge, they have to date not been implemented for online use.

### Front-end and back-end co-design

We have seen that preventative or front-end methods alone are not sufficient for full cancellation of stimulation artifacts. At the same time, adequate artifact prevention and front-end mitigation methods are required to enable back-end cancellation of sampled artifacts. For example, since subtractive algorithms require linearity of superimposed neural and artifact signals, they must be paired with high dynamic range front-ends to prevent signal loss. Therefore, back-end signal processing and resilient front-ends must be co-designed in closed-loop neuromodulation systems for sufficient artifact cancellation.

Two recently designed systems demonstrate this symbiotic combination of artifact cancellation methods. Culaclii *et al.* [37] demonstrate a system that implements template subtraction in two stages separated between the front-end and back-end. Front-end subtraction keeps the recorded signal within the linear range of the amplifiers without requiring gain reduction, and back-end subtraction cancels any remaining artifact. A full system for closed-loop neuromodulation developed by Zhou *et al.* [49\*] incorporates the recording and stimulation circuits presented by Johnson *et al.* [19\*\*] and on-board processing for back-end artifact cancellation. The stimulator utilizes preventative charge-balancing techniques, while the front-end also provides a high dynamic range and rapid recovery from artifacts that permits the use of linear interpolation for back-end cancellation, losing only one sample of information in the process. While high dynamic range is not strictly required, it aids the recording front-end to have rapid recovery from the artifact. To our knowledge, this work was the first demonstration of system-wide, on-line, hardware-based artifact cancellation *in vivo*.

### Conclusions

Closed-loop neuromodulation holds the promise of transforming disease treatment to be performed in an intelligent, patient-specific, and on-demand manner using devices that learn neurological biomarkers and



**Table 1****Comparison of techniques for artifact prevention and mitigation**

Methods	Variants	Toward closed-loop
<i>Prevention</i>		
Stimulation pulse charge-balance	Voltage offset correction [16] Current copying [17,18] Current source reuse [19**,20,21]	Improving charge balance reduces artifact size and duration, relaxing requirements on signal acquisition chain
Stimulation waveform design	Tri-phasic stimulation [22] Zero-forcing equalization of waveform [23]	Compensates for artifact-inducing properties of stimulator, neural tissue and recording circuitry
Electrode and reference configuration	Symmetric stim and sense electrode geometry [24**] Artifact-tracking voltage supply [25]	Keeps artifact common-mode, which can be tracked by the supply and also cancelled through differential amplification
<i>Front-end techniques</i>		
Saturation prevention	High Dynamic Range [19**,36] Front-end subtraction [32,37,38*,39] Electrode disconnection [1,40–43]	Keeps recorded artifacts linear, improving the performance of back-end techniques
Rapid recovery	Amplifier charge reset [19**,44] High-pass pole shifting [12*,45,56] Active electrode discharge [12*,13,14]	Recovers from saturation quickly, reduces data loss, and lowers requirements on front-end dynamic range
<i>Back-end techniques</i>		
Data reconstruction	Sample-and-hold interpolation (offline) [47,48] Linear interpolation (online) [49*] Linear interpolation (offline) [50*] Gaussian interpolation (offline) [51] Cubic spline interpolation (offline) [52]	Simplest to implement and is effective with relaxed SNR requirements
Artifact subtraction	Averaged template subtraction (online) [37,55*,56] Averaged template subtraction (offline) [48,53,54] Averaged template resampling and subtraction (offline) [57,58] Function fitting template subtraction (offline) [9,53,59] Adaptive filter (online) [38*]	Can theoretically remove artifact without distortion to underlying signal if paired with high-dynamic-range front-end
Component decomposition	Ensemble empirical mode decomposition (offline) [61,62] Independent component analysis (offline) [61,63]	Can remove stimulation artifact while providing other information and removing other types of artifact

automatically adapt stimulation for optimized therapies. To enable this vision, future devices will have hundreds to thousands (or more) of recording and stimulation channels with embedded signal processing and intelligence. There are numerous challenges in the realization of such devices, and here we have reviewed the state-of-the-art for addressing one key challenge: the ability to perform recording, stimulation, and computation simultaneously and unhindered by stimulation artifacts.

We have presented various artifact mitigation techniques at different points in the stimulation and signal path. **Table 1** summarizes these techniques and discusses their key contributions for integration into a closed-loop neuromodulation system. An optimized system must combine preventative, front-end, and back-end techniques, which, in combination, we believe are key to obtaining the best results in recording the full underlying neural signal during stimulation and thereby enabling true closed-loop neuromodulation.

Current recording front-ends with artifact mitigation have already been demonstrated in high density integrated

chips, consuming only a few microwatts and occupying less than 0.2 mm<sup>2</sup> per channel [19\*\*,38\*]. Back-end artifact cancellation algorithms have been demonstrated in both custom hardware and on-board software running in real time. Platform devices, combining front-end and back-end methods with algorithmic flexibility and reprogrammability, have already been miniaturized to weigh under 18 g with battery lives of over 11 hours [49\*], and monolithic integration of these components would result in extremely low power microsystems for improved implant safety and device lifetime. Still, scaling to hundreds or thousands of channels remains a challenge, since conventional front-ends that have achieved significant scaling often trade off artifact-resilient features for improved area. Further research in this area could yield neural interfaces that will completely prevent stimulation artifacts from occurring in the first place and allow for high-performance recording during stimulation without any loss of information.

### Conflict of interest statement

Nothing declared.

## Acknowledgements

This work was supported in part by the Defense Advanced Research Projects Agency (W911NF-14-2-0043), the National Science Foundation Graduate Research Fellowship Program (Grant No. 1106400), the Chan-Zuckerberg Biohub, and the sponsors of the Berkeley Wireless Research Center. The authors thank Profs. Jan Rabaey, Jose Carmena, Elad Alon, and Dr. Samantha Santacruz.

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